

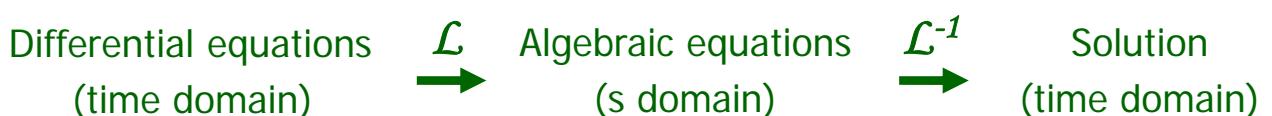


Chap 3 State Variable Models

林沛群
國立台灣大學
機械工程學系

Background

- Classic control theory – transfer function approach
 - ◆ For linear-time-invariant (LTI) and single-input-single-output (SISO) systems

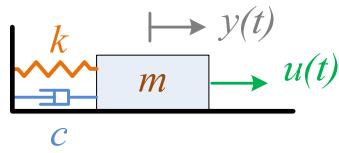


- Modern control theory – state-space approach
 - ◆ Deal with differential equations directly
 - ◆ Can be utilized for time-variant, nonlinear, and multiple-input-multiple-output (MIMO) systems
 - ◆ 線性控制 為其他進階控制課程的基礎

Five Representations of a System -1

以SMD system為例

$$\begin{aligned} \text{input} &= u \\ \text{output} &= y \end{aligned}$$



- (1) Differential equation

$$m\ddot{y} + c\dot{y} + ky = u$$

- (2) Transfer function

$$G(s) = \frac{Y(s)}{U(s)} = \frac{1}{ms^2 + cs + k}$$

note I.C. s = 0

- (3) Impulse response

$$u(t) = \delta(t) \xrightarrow{\mathcal{L}} U(s) = 1$$

$$Y(s) = G(s)U(s) = \frac{1}{ms^2 + cs + k}$$

Five Representations of a System -2

- (4) State-space

$$\text{output} = y = x_1$$

$$\dot{y} = x_2 = \dot{x}_1$$

$$\ddot{y} = \dot{x}_2 = \frac{-c}{m}\dot{y} + \frac{-k}{m}y + \frac{1}{m}u = \frac{-c}{m}x_2 + \frac{-k}{m}x_1 + \frac{1}{m}u$$

$$\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}_{2 \times 1} = \begin{bmatrix} 0 & 1 \\ -\frac{k}{m} & -\frac{c}{m} \end{bmatrix}_{2 \times 2} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}_{2 \times 1} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix}_{2 \times 1} u_{1 \times 1}$$

Note:
A 3rd-order system
2 inputs
2 outputs

$$y = \begin{bmatrix} 1 & 0 \end{bmatrix}_{1 \times 2} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}_{2 \times 1} + \begin{bmatrix} 0 \end{bmatrix}_{1 \times 1} u_{1 \times 1}$$

x: 3x1
u: 2x1
y: 2x1

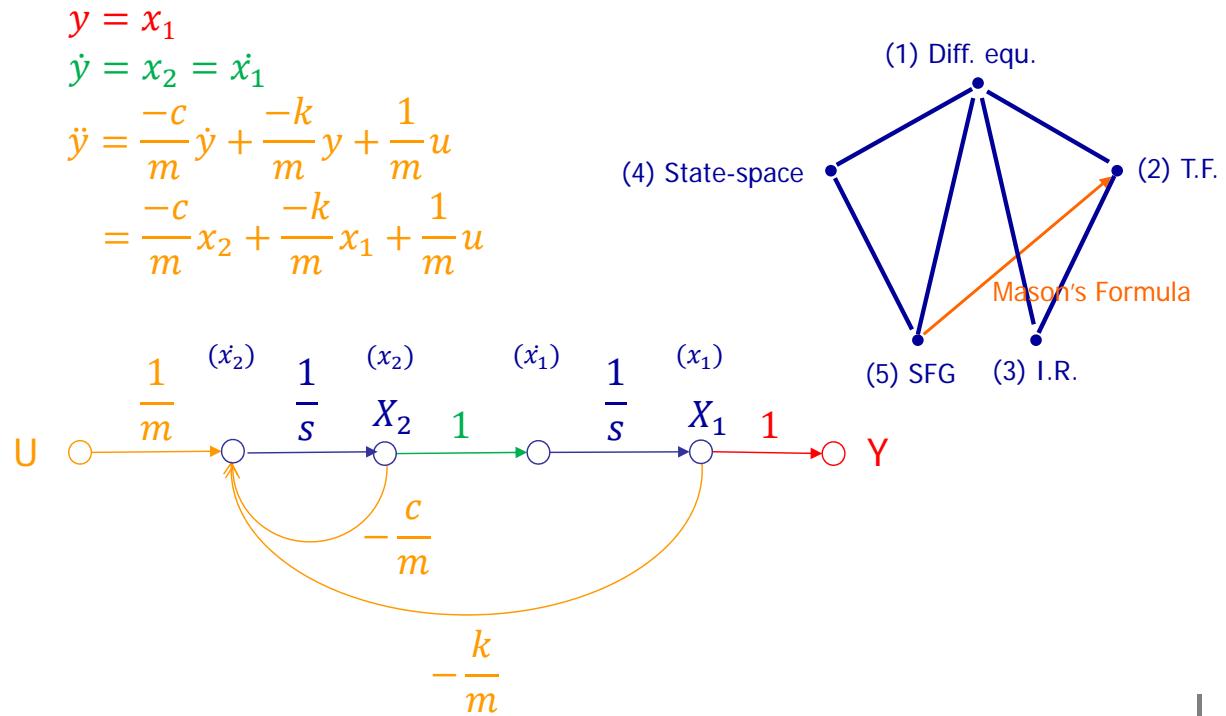
$$\begin{array}{l} \dot{x} = Ax + Bu \\ y = Cx + Du \end{array} \quad \begin{bmatrix} A & B \\ C & D \end{bmatrix}$$

A: 3x3
B: 3x2
C: 2x3
D: 2x2

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix}_{5 \times 5}$$

Five Representations of a System -3

□ (5) Signal-flow graph



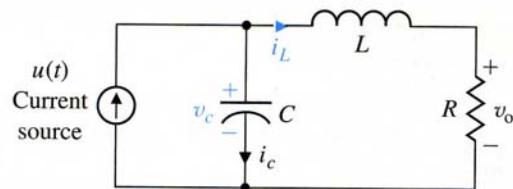
自動控制 ME3007-01 Chap 3 - 林沛群

5

Five Representations of a System -4

→ 以 RLC circuit 為例

$$\begin{aligned} \text{input} &= u \\ \text{output} &= v_o \end{aligned}$$



□ (1) Differential equation

$$i_c = C \frac{dv_c}{dt} = u(t) - i_L \quad \dots\dots (1)$$

$$L \frac{di_L}{dt} = v_c - i_L R \quad \dots\dots (2)$$

$$v_o = i_L R \quad \dots\dots (3)$$

自動控制 ME3007-01 Chap 3 - 林沛群

6

Five Representations of a System -5

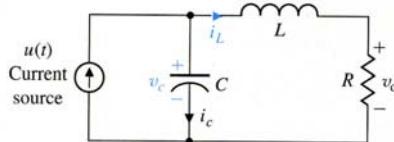
□ (2) State-space

Set $x_1 = v_c$ $x_2 = i_L$

$$C \frac{dv_c}{dt} = u(t) - i_L$$

$$L \frac{di_L}{dt} = v_c - i_L R$$

$$v_o = i_L R$$

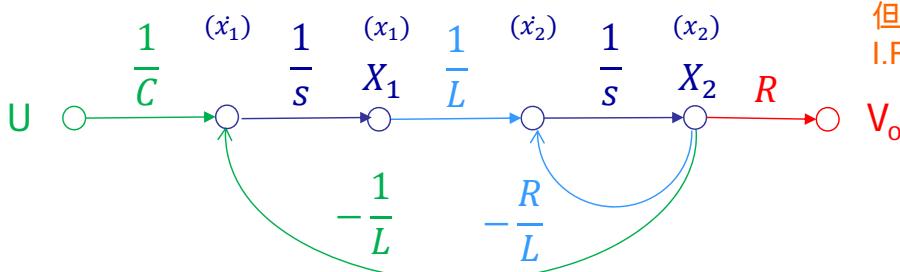


$$\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} 0 & -\frac{1}{C} \\ \frac{1}{L} & -\frac{R}{L} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ \frac{1}{C} \end{bmatrix} u$$

針對一個系統，
選擇不同state，就會
有不同的(A,B,C,D)和
SFG。

$$y = [0 \quad R] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

□ (3) Signal-flow graph



但Transfer function和
I.R.則為唯一表達

7

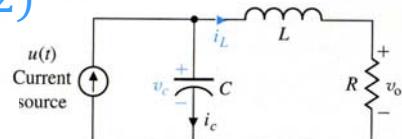
自動控制 ME3007-01 Chap 3 - 林沛群

Five Representations of a System -6

□ (4) Transfer function

$$\begin{aligned} \text{From (1)} \quad u(t) &= i_c + i_L = C \frac{dv_c}{dt} + i_L & (2) \\ &= C \frac{d}{dt} \left(L \frac{di_L}{dt} + i_L R \right) + i_L & (2) \\ &= CL \ddot{i}_L + CR \dot{i}_L + i_L & (3) \\ &= \frac{CL}{R} \ddot{v}_o + C \dot{v}_o + \frac{v_o}{R} & (3) \end{aligned}$$

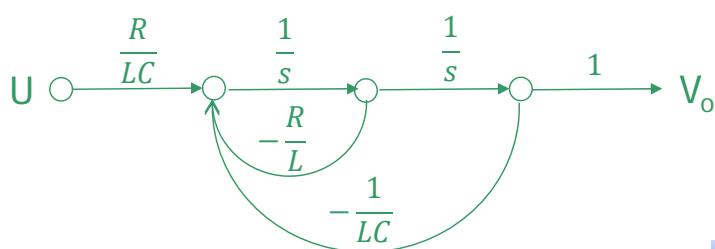
$$\rightarrow G = \frac{V_o}{U} = \frac{1}{\frac{LC}{R}s^2 + Cs + \frac{1}{R}}$$



仿SMD example中
state架構所得出之SFG

□ (5) Impulse response

$$V_o = GU = G$$



自動控制 ME3007-01 Chap 3 - 林沛群

8

Solving Differential equations -1

□ Single variable

$$\dot{x} = ax + bu \xrightarrow{\mathcal{L}} sX(s) - x(0) = aX(s) + bU(s)$$

$$X(s) = \frac{x(0)}{s-a} + \frac{b}{s-a} U(s)$$
$$\downarrow \mathcal{L}^{-1}$$
$$x(t) = e^{at}x(0) + \int_0^t e^{a(t-\tau)}bu(\tau)d\tau$$

Note: $\mathcal{L}^{-1}[F(s)G(s)] = f(t) * g(t)$

Solving Differential equations -2

□ Multiple variables

$$\begin{matrix} \dot{x} \\ \text{nx1} \end{matrix} = \begin{matrix} A \\ \text{nxn} \end{matrix} \begin{matrix} x \\ \text{nx1} \end{matrix} + \begin{matrix} B \\ \text{nxm} \end{matrix} \begin{matrix} u \\ \text{mx1} \end{matrix} \xrightarrow{\mathcal{L}} sX(s) - x(0) = AX(s) + BU(s)$$

$$(sI - A)X(s) = x(0) + BU(s)$$

$$\begin{aligned} X(s) &= (sI - A)^{-1}x(0) + (sI - A)^{-1}BU(s) \\ &= \Phi(s)x(0) + \Phi(s)BU(s) \end{aligned}$$

$$\downarrow \mathcal{L}^{-1}$$

State transition matrix

$$\begin{array}{ccc} \Phi(t) & \xleftarrow[\mathcal{L}^{-1}]{\mathcal{L}} & \Phi(s) \\ \parallel & & \parallel \\ e^{At} & & (sI - A)^{-1} \\ \parallel \\ I + At + A^2 \frac{t^2}{2!} + A^3 \frac{t^3}{3!} + \dots \end{array}$$

$$\begin{aligned} x(t) &= \Phi(t)x(0) + \int_0^t \Phi(t-\tau)Bu(\tau)d\tau \\ &= e^{At}x(0) + \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau \end{aligned}$$

State-space to Transfer Function -1

□ $\dot{x} = Ax + Bu$

$$y = Cx + Du$$

由前一頁公式

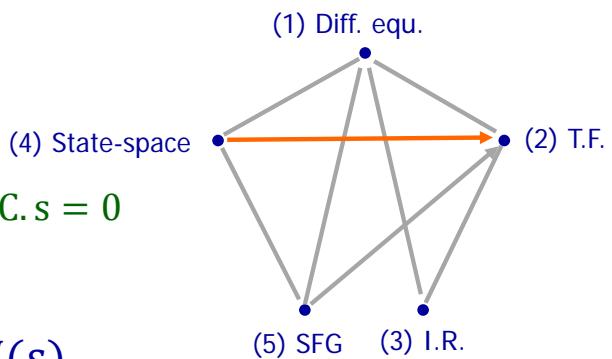
$$X(s) = (sI - A)^{-1}BU(s) \quad \text{note I.C. } s = 0$$

$$Y(s) = CX(s) + DU(s)$$

$$= C(sI - A)^{-1}BU(s) + DU(s)$$

$$= [C(sI - A)^{-1}B + D]U(s)$$

$$= G(s)U(s)$$



➡ $G(s) = C(sI - A)^{-1}B + D$

State-space to Transfer Function -2

□ Revisit the RLC circuit

$$A = \begin{bmatrix} 0 & -\frac{1}{C} \\ 1 & -\frac{R}{L} \end{bmatrix} \quad B = \begin{bmatrix} \frac{1}{C} \\ 0 \end{bmatrix}$$

$$C = [0 \quad R] \quad D = 0$$

Note: $M = \begin{bmatrix} a & b \\ c & d \end{bmatrix}$

$$M^{-1} = \frac{1}{ad - bc} \begin{bmatrix} d & -b \\ -c & a \end{bmatrix}$$

$$sI - A = \begin{bmatrix} s & \frac{1}{C} \\ -\frac{1}{L} & s + \frac{R}{L} \end{bmatrix} \quad MM^{-1} = I$$

$$\Phi(s) = (sI - A)^{-1} = \frac{1}{s^2 + \frac{R}{L}s + \frac{1}{LC}} \begin{bmatrix} s + \frac{R}{L} & -\frac{1}{C} \\ \frac{1}{L} & s \end{bmatrix}$$

$$G(s) = [0 \quad R]\Phi(s) \begin{bmatrix} 1 \\ \frac{1}{C} \\ 0 \end{bmatrix} = \frac{\frac{R}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

和p8結果相同

Evaluation of the State Transition Matrix -1

- Revisit the RLC circuit

$$R = 3 \ L = 1 \ C = \frac{1}{2} \quad \therefore A = \begin{bmatrix} 0 & -2 \\ 1 & -3 \end{bmatrix}$$

$$\Phi(s) = (sI - A)^{-1} = \frac{1}{(s+1)(s+2)} \begin{bmatrix} s+3 & -2 \\ 1 & s \end{bmatrix}$$

$$\Phi_{11} = \frac{(s+3)}{(s+1)(s+2)} = \frac{2}{s+1} + \frac{-1}{s+2}$$

同理可得 Φ_{12} Φ_{21} Φ_{22}

$$\Phi(t) = \mathcal{L}^{-1}[\Phi(s)] = \begin{bmatrix} 2e^{-t} - e^{-2t} & -2e^{-t} + 2e^{-2t} \\ e^{-t} - e^{-2t} & -e^{-t} + 2e^{-2t} \end{bmatrix}$$

Evaluation of the State Transition Matrix -2

- Revisit the RLC circuit

Another method: *eigen decomposition*

$$AV = V\Lambda \quad \begin{array}{l} \text{eigenvectors} \\ \text{eigenvalues} \end{array}$$
$$\begin{bmatrix} 0 & -2 \\ 1 & -3 \end{bmatrix} \begin{bmatrix} \frac{2}{\sqrt{5}} & \frac{1}{\sqrt{2}} \\ \frac{1}{\sqrt{5}} & \frac{1}{\sqrt{2}} \end{bmatrix} = \begin{bmatrix} \frac{2}{\sqrt{5}} & \frac{1}{\sqrt{2}} \\ \frac{1}{\sqrt{5}} & \frac{1}{\sqrt{2}} \end{bmatrix} \begin{bmatrix} -1 & 0 \\ 0 & -2 \end{bmatrix}$$

$$A = V\Lambda V^{-1}$$

$$\begin{aligned} e^{At} &= Ve^{\Lambda t}V^{-1} \\ &= \begin{bmatrix} 2 & 1 \\ 1 & 1 \end{bmatrix} \begin{bmatrix} e^{-t} & 0 \\ 0 & e^{-2t} \end{bmatrix} \frac{1}{\sqrt{5}} \begin{bmatrix} 1 & -1 \\ -1 & 2 \end{bmatrix} \\ &= \begin{bmatrix} 2e^{-t} - e^{-2t} & -2e^{-t} + 2e^{-2t} \\ e^{-t} - e^{-2t} & -e^{-t} + 2e^{-2t} \end{bmatrix} \end{aligned}$$

和前頁p13結果相同

Transfer Function to State-space -1

□ Ex: $G(s) = \frac{b_3s^3 + b_2s^2 + b_1s + b_0}{s^4 + a_3s^3 + a_2s^2 + a_1s + a_0}$

$$= \frac{b_3s^3 + b_2s^2 + b_1s + b_0}{s^4 + a_3s^3 + a_2s^2 + a_1s + a_0} \frac{Z(s)}{Z(s)} = \frac{Y(s)}{U(s)}$$

$$Y(s) = (b_3s^3 + b_2s^2 + b_1s + b_0)Z(s)$$

$$y(t) = b_3z^{(3)} + b_2\ddot{z} + b_1\dot{z} + b_0z$$

$$U(s) = (s^4 + a_3s^3 + a_2s^2 + a_1s + a_0)Z(s)$$

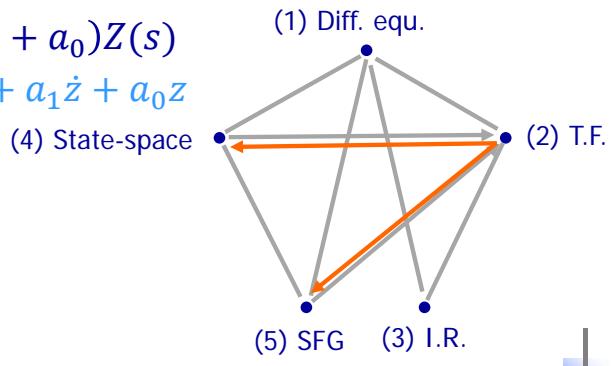
$$u(t) = z^{(4)} + a_3z^{(3)} + a_2\ddot{z} + a_1\dot{z} + a_0z$$

Assign $x_1 = z$

$$x_2 = \dot{z} = \dot{x}_1$$

$$x_3 = \ddot{z} = \dot{x}_2$$

$$x_4 = z^{(4)} = \dot{x}_3$$

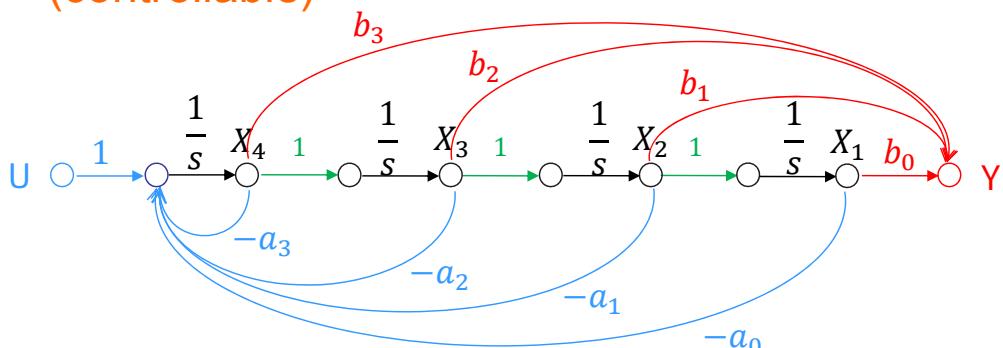


Transfer Function to State-space -2

□ $\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -a_0 & -a_1 & -a_2 & -a_3 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \end{bmatrix} u$

$$y = [b_0 \ b_1 \ b_2 \ b_3] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix}$$

Phase variable canonical form
(controllable)

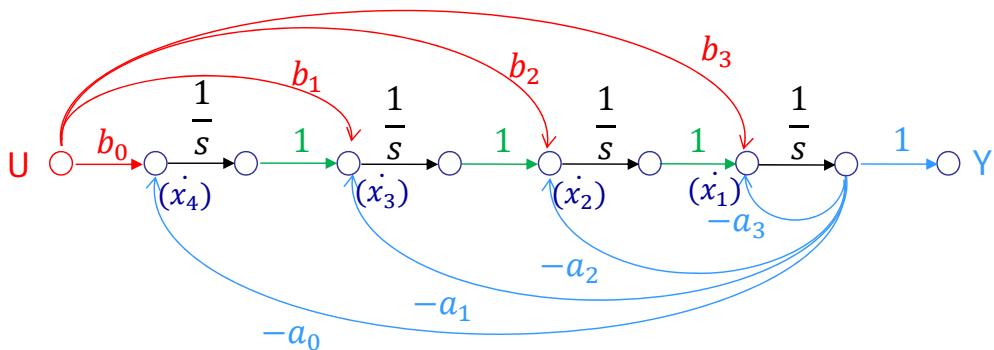


Transfer Function to State-space -3

□ $\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} = \begin{bmatrix} -a_3 & 1 & 0 & 0 \\ -a_2 & 0 & 1 & 0 \\ -a_1 & 0 & 0 & 1 \\ -a_0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} b_3 \\ b_2 \\ b_1 \\ b_0 \end{bmatrix} u$

$$y = [1 \ 0 \ 0 \ 0] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix}$$

Input forward canonical form
(observable)



自動控制 ME3007-01 Chap 3 - 林沛群

17

終

□ Questions?

